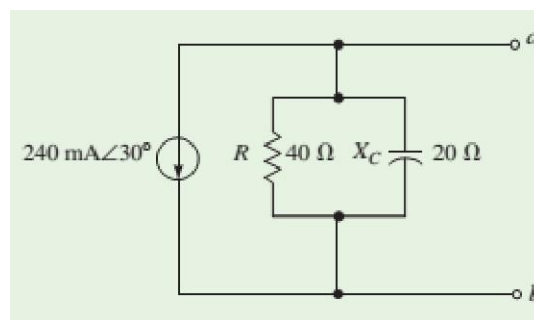




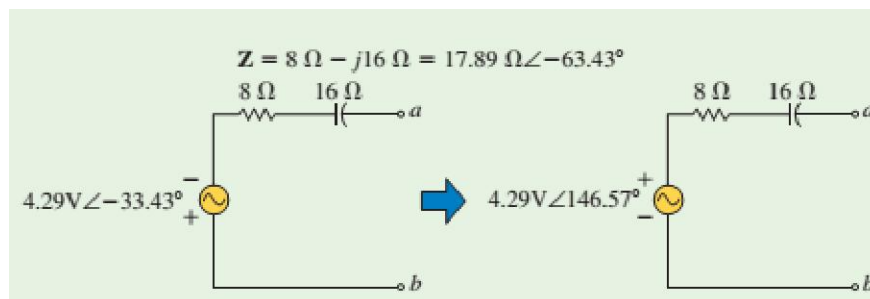
## Sheet (1) solution... Review

1. Convert the current source of Figure (1) into an equivalent voltage source.



$$\begin{aligned} Z &= \frac{(40 \Omega \angle 0^\circ)(20 \Omega \angle -90^\circ)}{40 \Omega - j20 \Omega} \\ &= \frac{800 \Omega \angle -90^\circ}{44.72 \angle -26.57^\circ} \\ &= 17.89 \Omega \angle -63.43^\circ = 8 \Omega - j16 \Omega \end{aligned}$$

$$\begin{aligned} E &= (240 \text{ mA} \angle 30^\circ)(17.89 \Omega \angle -63.43^\circ) \\ &= 4.29 \text{ V} \angle -33.43^\circ \end{aligned}$$



2. Given the circuit of Figure (2), write the loop equations and solve for the loop currents.

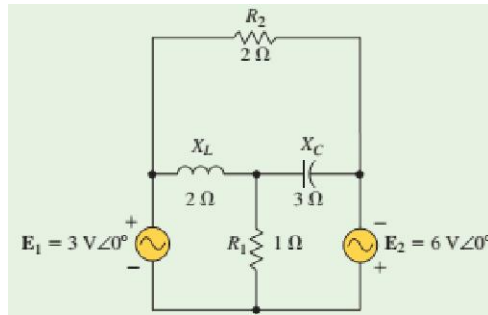


Figure (2)

The loop equations may now be written as

Loop 1:  $(Z_1 + Z_3)I_1 - (Z_3)I_2 - (Z_1)I_3 = E_1$   
 Loop 2:  $-(Z_3)I_1 + (Z_2 + Z_3 + Z_4)I_2 - (Z_4)I_3 = 0$   
 Loop 3:  $-(Z_1)I_1 - (Z_4)I_2 + (Z_1 + Z_4)I_3 = E_2$

Using the given impedance values, we have

Loop 1:  $(1 \Omega + j2 \Omega)I_1 - (j2 \Omega)I_2 - (1 \Omega)I_3 = 3 \text{ V}$   
 Loop 2:  $-(j2 \Omega)I_1 + (2 \Omega - j1 \Omega)I_2 - (-j3 \Omega)I_3 = 0$   
 Loop 3:  $-(1 \Omega)I_1 - (-j3 \Omega)I_2 + (1 \Omega - j3 \Omega)I_3 = 6 \text{ V}$

Notice that in the preceding equations, the phase angles ( $\theta = 0^\circ$ ) for the voltages have been omitted. This is because  $3 \text{ V} \angle 0^\circ = 3 \text{ V} + j0 \text{ V} = 3 \text{ V}$ .

Using a calculator such as the Sharp EL-506W, we have:

$I_1 = 5.26 \text{ A} \angle -41.25^\circ$   
 $I_2 = 3.18 \text{ A} \angle -45^\circ$   
 $I_3 = 3.39 \text{ A} \angle -1.03^\circ$

3. Use nodal analysis to determine the voltage  $V$  for the circuit of Figure (3).

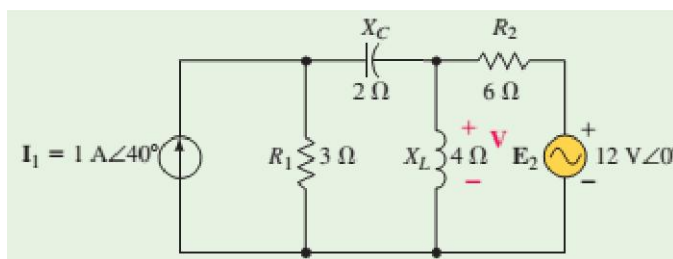
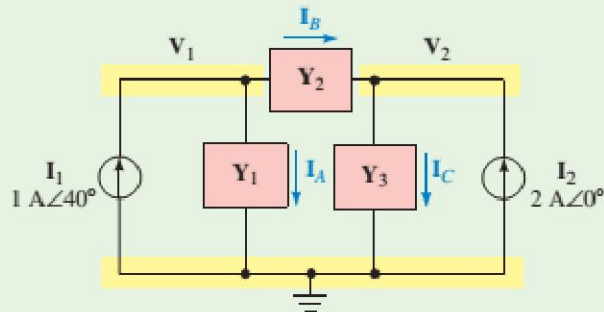
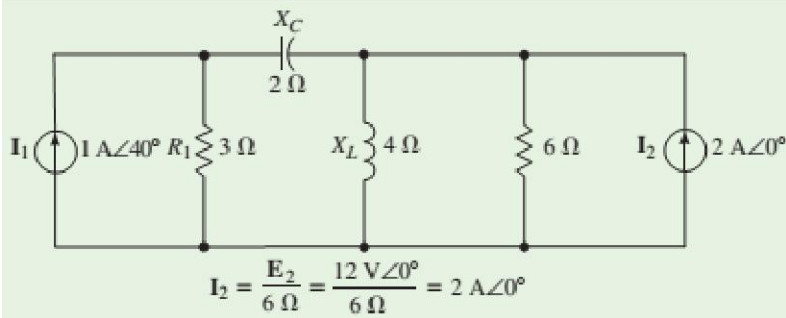


Figure (3)



$$Y_1 = \frac{1}{3} S + j0$$

$$Y_2 = 0 + j\frac{1}{2} S$$

$$Y_3 = \frac{1}{6} S - j\frac{1}{4} S$$

Node 1:  $I_A + I_B = I_1$

$$Y_1 V_1 + Y_2 (V_1 - V_2) = I_1$$

$$(Y_1 + Y_2) V_1 - Y_2 V_2 = I_1$$

Node 2:  $I_C = I_B + I_2$

$$Y_3 V_2 = Y_2 (V_1 - V_2) + I_2$$

$$-Y_2 V_1 + (Y_2 + Y_3) V_2 = I_2$$

Now, substituting the values of admittance into the nodal equations, we get

Node 1:  $\left(\frac{1}{3} + \frac{1}{-j2}\right) V_1 - \left(\frac{1}{-j2}\right) V_2 = 1 \angle 40^\circ$

Node 2:  $-\left(\frac{1}{-j2}\right) V_1 + \left(\frac{1}{-j2} + \frac{1}{j4} + \frac{1}{6}\right) V_2 = 2 \angle 0^\circ$

Node 1:  $\left(\frac{1}{3} + j\frac{1}{2}\right) V_1 - \left(j\frac{1}{2}\right) V_2 = 1 \angle 40^\circ$

Node 2:  $-\left(j\frac{1}{2}\right) V_1 + \left(\frac{1}{6} + j\frac{1}{4}\right) V_2 = 2 \angle 0^\circ$

$$V_1 = 5.29 \text{ V} \angle 48.75^\circ$$

$$V_2 = 5.80 \text{ V} \angle 33.27^\circ$$

4. Determine the Y equivalent of the  $\Delta$  network shown in Figure (4), show how to redistribute the  $\Delta$  network to return in the form of Y.

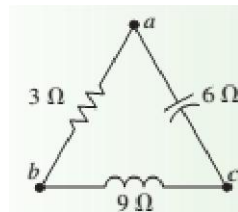


Figure (4)

$$Z_1 = \frac{(3 \Omega)(-j6 \Omega)}{3 \Omega - j6 \Omega + j9 \Omega} = \frac{-j18 \Omega}{3 + j3} = \frac{18 \Omega \angle -90^\circ}{4.242 \angle 45^\circ}$$

$$= 4.242 \Omega \angle -135^\circ$$

$$= -3.0 \Omega - j3.0 \Omega$$

$$Z_2 = \frac{(3 \Omega)(j9 \Omega)}{3 \Omega - j6 \Omega + j9 \Omega} = \frac{j27 \Omega}{3 + j3} = \frac{27 \Omega \angle 90^\circ}{4.242 \angle 45^\circ}$$

$$= 6.364 \Omega \angle 45^\circ$$

$$= 4.5 \Omega + j4.5 \Omega$$

$$Z_3 = \frac{(j9 \Omega)(-j6 \Omega)}{3 \Omega - j6 \Omega + j9 \Omega} = \frac{54 \Omega}{3 + j3} = \frac{54 \Omega \angle 0^\circ}{4.242 \angle 45^\circ}$$

$$= 12.73 \Omega \angle -45^\circ$$

$$= 9.0 \Omega - j9.0 \Omega$$

The second requirement ... self study

5. Consider the circuit of Figure (5), Find  $V_R$  and  $V_C$  using the superposition theorem.

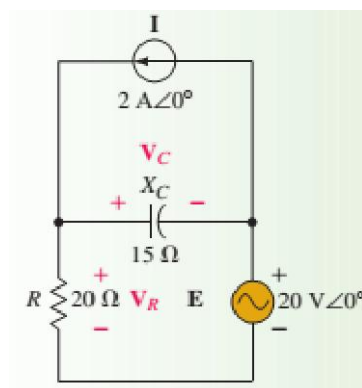


Figure (5)

The superposition theorem may be employed as follows:

**Voltages due to the current source:** Eliminating the voltage source, we obtain the circuit shown in Figure

The impedance “seen” by the current source will be the parallel combination of  $R \parallel Z_C$ .

$$Z_1 = \frac{(20 \Omega)(-j15 \Omega)}{20 \Omega - j15 \Omega} = \frac{300 \Omega \angle -90^\circ}{25 \Omega \angle -36.87^\circ} = 12 \Omega \angle -53.13^\circ$$

The voltage  $V_{R(1)}$  is the same as the voltage across the capacitor,  $V_{C(1)}$ . Hence,

$$\begin{aligned} V_{R(1)} &= (2 \text{ A} \angle 0^\circ)(12 \Omega \angle -53.13^\circ) \\ &= 24 \text{ V} \angle -53.13^\circ \end{aligned}$$

**Voltages due to the voltage source:** Eliminating the current source, we have the circuit shown in Figure

The voltages  $V_{R(2)}$  and  $V_{C(2)}$  are determined by applying the voltage divider rule,

$$\begin{aligned} V_{R(2)} &= \frac{20 \Omega \angle 0^\circ}{20 \Omega - j15 \Omega} (20 \text{ V} \angle 0^\circ) \\ &= \frac{400 \text{ V} \angle 0^\circ}{25 \angle -36.87^\circ} = 16 \text{ V} \angle +36.87^\circ \end{aligned}$$

$$\begin{aligned} V_R &= V_{R(1)} + V_{R(2)} \\ &= 24 \text{ V} \angle -53.13^\circ + 16 \text{ V} \angle 36.87^\circ \\ &= (14.4 \text{ V} - j19.2 \text{ V}) + (12.8 \text{ V} + j9.6 \text{ V}) \\ &= 27.2 \text{ V} - j9.6 \text{ V} \\ &= 28.84 \text{ V} \angle -19.44^\circ \end{aligned}$$

6. Determine the Thévenin equivalent circuit external to  $Z_L$  in the circuit in Figure (6)

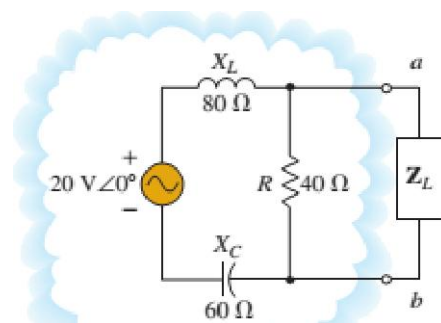
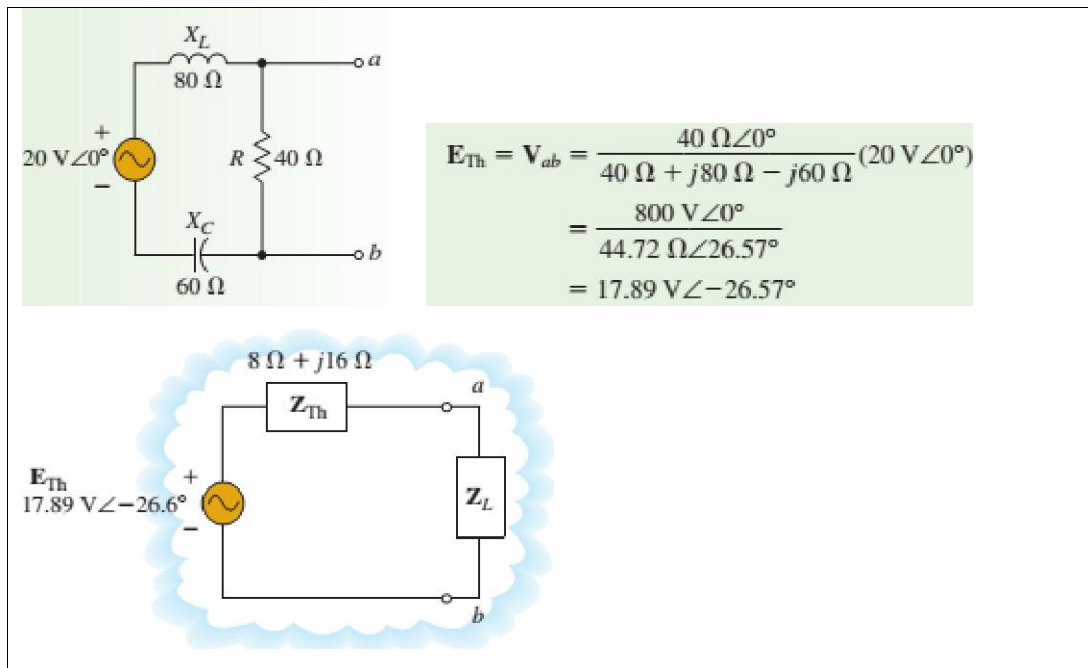


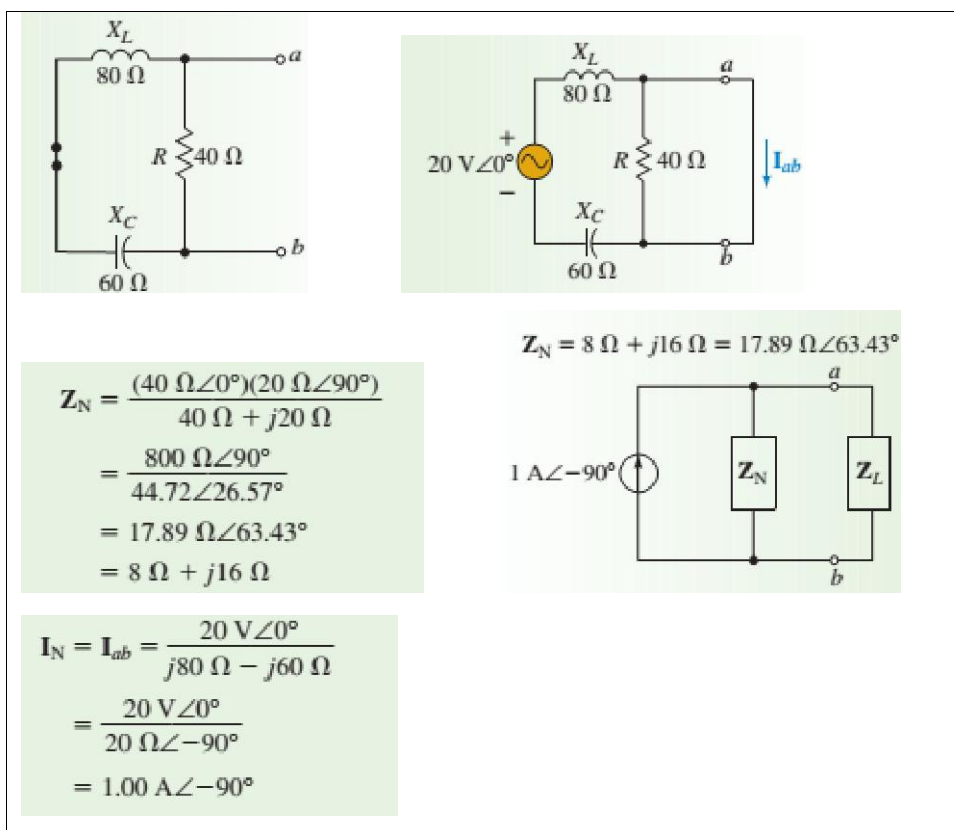
Figure (6)

$$\begin{aligned} Z_{Th} &= R \parallel (Z_L + Z_C) \\ &= \frac{(40 \Omega \angle 0^\circ)(20 \Omega \angle 90^\circ)}{40 \Omega + j20 \Omega} \\ &= \frac{800 \Omega \angle 90^\circ}{44.72 \Omega \angle 26.57^\circ} \\ &= 17.89 \Omega \angle 63.43^\circ \\ &= 8 \Omega + j16 \Omega \end{aligned}$$

Voltage source is replaced with a short circuit.



7. Repeat the previous problem but by using the Norton equivalent.



*Good Luck*

Dr. Basem ElHalawany